

**Amendments to the Claims:**

This listing of claims will replace all prior versions, and listings, of claims in the application:

**Listing of Claims:**

Claim 1 (currently amended): A method for converting a digital input value (~~Sq1~~) quantized according to a first quantization coefficient (~~Cq1~~) and encoded over and most n1 bits, into a digital output value (~~Sq2~~) quantized according to a second quantization coefficient (~~Cq2~~) and encoded over and most n2 bits, where n1 and n2 are nonzero integers, comprising the steps ~~consisting in~~ of:

a) ~~multiplying the digital input value~~(~~Sq1~~) by an integer B encoded over at most  $\beta$  bits, where  $\beta$  is a nonzero integer, ~~in order so as~~ to generate a first intermediate digital value (~~C~~) encoded over at most  $n1+\beta$  bits; and

b) ~~fixed-point dividing said first intermediate digital value~~ (~~C~~) by the number  $2^\alpha$ , where  $\alpha$  is an integer less than or equal to  $n1+\beta$ , ~~in order so as~~ to generate said digital output value (~~Sq2~~),

wherein the number  $\frac{B}{2^\alpha}$  is substantially equal to the ratio of said second quantization coefficient (~~Cq2~~) to said first quantization coefficient(~~Cq1~~);

and wherein ~~step b)~~ the step of fixed point dividing is carried out by means of a sigma-delta modulator.

Claim 2 (currently amended): The method as claimed in claim 1, wherein ~~step b)~~ the step of fixed-point dividing comprises the steps ~~consisting in~~ of:

b1) ~~adding said first intermediate digital value~~(~~C~~), on the one hand, and a digital error value (~~E~~) encoded over at most  $\alpha$  bits, on the other hand, ~~in order so as~~ to generate a second intermediate digital value (~~D~~) encoded over at most  $n1+\beta+1$  bits;

~~b2)~~ selecting ~~the  $n2$~~  a given number  $n2$  of the most significant bits of said second intermediate digital value ( $\mathcal{D}$ ) as the digital output value ( $\mathcal{S}q2$ ), where  $n2$  is equal to  $n1+\beta+1-\alpha$ , and

~~b3)~~ selecting ~~the  $\alpha$~~  a given number  $\alpha$  of the least significant bits of said second intermediate digital value ( $\mathcal{D}$ ) as the digital error value ( $\mathcal{E}$ ).

Claim 3 (currently amended): The method as claimed in claim 2, wherein ~~step b2) and step b3)~~ the steps of selecting are carried out together with the aid of a discriminator for separating said the  $n1+\beta+1-\alpha$  most significant bits of said the second intermediate digital value ( $\mathcal{D}$ ), on the one hand, and ~~said the~~  $\alpha$  least significant bits of said second intermediate digital value ( $\mathcal{D}$ ), on the other hand.

Claim 4 (currently amended): The method as claimed in claim 2, wherein ~~step b2)~~ the step of selecting the  $n2$  most significant bits is carried out via an operation of shifting to the right by  $\alpha$  bits, which is applied to the  $n1+\beta+1$  bits of the second intermediate digital value ( $\mathcal{D}$ ).

Claim 5 (currently amended): The method as claimed in claim 4, wherein ~~step b3)~~ the step of selecting the  $\alpha$  least significant bits is carried out by applying to the second intermediate digital value ( $\mathcal{D}$ ) a mask having at most  $n1+\beta+1$  bits, the  $n1+\beta+1-\alpha$  most significant bits of which are equal to the logical value 0 and the  $\alpha$  least significant bits of which are equal to the logical value 1.

Claim 6 (currently amended): The method as claimed in claim 4, wherein ~~step b3)~~ the step of selecting the  $\alpha$  least significant bits is carried out, on the one hand, by an operation of shifting to the left by  $\alpha$ , which is applied to the  $n1+\beta+1-\alpha$  bits of the digital output value ( $\mathcal{S}q2$ ) for

generating a third intermediate digital value (~~F~~) encoded over at most  $n1+\beta+1$  bits and, on the other hand, by a difference operation between said third intermediate digital value (~~F~~) and said first intermediate digital value (~~G~~).

Claim 7 (currently amended): The method as claimed in ~~any one of the preceding claims~~ claim 1, wherein neither the first quantization coefficient nor the second quantization coefficient is an integer multiple of the other.

Claim 8 (currently amended): A device for converting a digital input value (~~Sq1~~) quantized according to a first quantization coefficient (~~Cq1~~) and encoded over at most  $n1$  bits, into a digital output value (~~Sq2~~) quantized according to a second quantization coefficient (~~Cq2~~) and encoded over at most  $n2$  bits, where  $n1$  and  $n2$  are nonzero integers, comprising:

- multiplier means ~~(10)~~ for multiplying the digital input value (~~Sq1~~) by an integer  $B$  encoded over at most  $\beta$  bits, where  $\beta$  is a nonzero integer, ~~generating so as to~~ generate a first intermediate digital value (~~G~~) encoded over at most  $n1+\beta$  bits; ~~and~~
- divider means for fixed-point dividing said first intermediate digital value (~~G~~) by the number  $2^\alpha$ , where  $\alpha$  is an integer less than or equal to  $n1+\beta$ , ~~generating so as to~~ generate said digital output value (~~Sq2~~),

$\frac{B}{2^\alpha}$

wherein the number  $\frac{B}{2^\alpha}$  is substantially equal to the ratio of said second quantization coefficient (~~Cq2~~) to said first quantization coefficient (~~Cq1~~);

and wherein said divider means comprise a sigma-delta modulator (~~20~~).

Claim 9 (currently amended): The device as claimed in claim 8, wherein the sigma-delta modulator (~~20~~) is a 1<sup>st</sup> order to sigma-delta modulator.

Claim 10 (currently amended): The device as claimed in claim 9, wherein the sigma-delta modulator (20) comprises:

- adder means (21) which receive as input said first intermediate digital value (C) as a first operand, on the one hand, and a digital error value (E) encoded over at most  $\alpha$  bits as a second operand, on the other hand, and which deliver as output a second intermediate digital value (D) encoded over at most  $n1+\beta+1$  bits;
- selection means (23) for selecting ~~the n2~~ a given number n2 of the most significant bits of said second intermediate digital value (D) as the digital output value (Sq2), where n2 is equal to  $n1+\beta+1-\alpha$  and for selecting the  $[[\alpha]]$  a given number  $\alpha$  of the least significant bits of said second intermediate digital value (D) as the digital error value (E).

Claim 11 (currently amended): The device as claimed in claim 10, wherein said selection means (23) ~~consist of~~ comprise a discriminator for separating ~~said the~~ n1+ $\beta$ +1- $\alpha$  most significant bits of ~~said the~~ the second intermediate digital value (D), on the one hand, and ~~said the~~ the  $\alpha$  least significant bits of said second intermediate digital value (D), on the other hand.

Claim 12 (currently amended): The device as claimed in claim 10, wherein said selection means (23) comprise an operator (24) for shifting to the right by  $\alpha$  bits, which receives as input the  $n1+\beta+1$  bits of the second intermediate digital value (D), and which delivers as output the  $n1+\beta+1-\alpha$  most significant bits of the second intermediate digital value (D) as a digital output value (Sq2).

Claim 13 (currently amended): The device as claimed in claim 12, wherein said selection means (23) further comprise means (25) for applying to the second intermediate digital value

~~(D)~~ a mask ~~(M)~~ having at most  $n1+\beta+1$  bits, the  $n1+\beta+1-\alpha$  most significant bits of which are equal to the logical value 0 and the  $\alpha$  least significant bits of which are equal to the logical value 1, so as to select the  $\alpha$  least significant bits of said second intermediate digital value ~~(D)~~ as the digital error value ~~(E)~~.

Claim 14 (currently amended): The device as claimed in claim 12, wherein said selection means ~~(23)~~ further comprise, on the one hand, an operator for shifting to the left by  $\alpha$  bits, which receives as input the  $n1+\beta+1-\alpha$  bits of the digital output value ~~(Seq2)~~ and delivers as output a third intermediate digital value ~~(F)~~ encoded over at most  $n1+\beta+1$  bits and, on the other hand, a difference operator which receives said third intermediate digital value ~~(F)~~ as a first operand and said first intermediate digital value ~~(C)~~ as a second operand, and which delivers as output ~~said the~~ digital error value ~~(E)~~.

Claim 15 (currently amended): The device as claimed in ~~any one of claims 10 to 14~~ claim 1, wherein the error signal ~~(E)~~ is delivered to the input of the adder means ~~(21)~~ through a unitary delay operator ~~(22)~~.

Claim 16 (currently amended): A digitally modulated frequency synthesizer, comprising a phase-locked loop ~~(PLL)~~ comprising a variable-ratio frequency divider ~~(14)~~ in the return path, wherein the division ratio is controlled by a digital value ~~(Se)~~ obtained in particular from a real value  $[(F_{ch})]$  corresponding to the central frequency of a radio channel, the synthesizer further comprising a conversion device ~~(18)~~ as claimed in ~~any one of claims 8 to 15~~ claim 8 for reducing the quantization error affecting said real value.